

How to Fool an ADC

Or how to hide the destruction of a turbine with the help of DSP

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Speakers Details

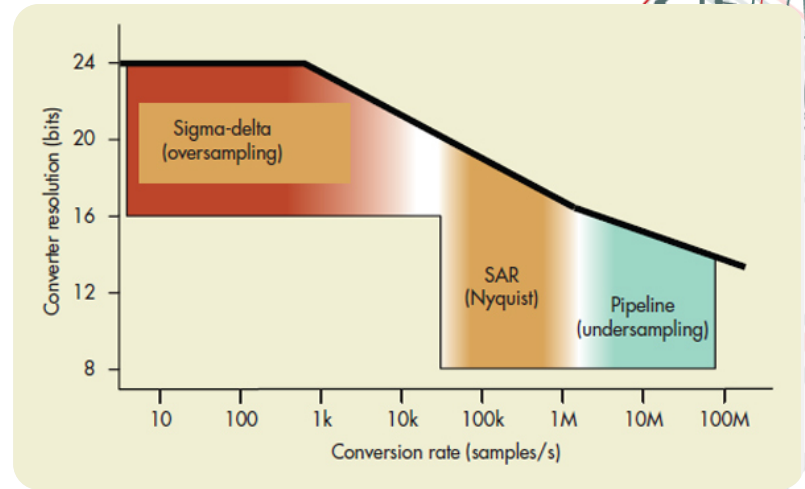
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Introduction to ADC

- A device that converts a continuous physical quantity (usually voltage) to a digital number that represents the quantity's amplitude
- An ADC is defined by its bandwidth and its signal to noise ratio
- Bandwidth of an ADC is characterized primarily by its sampling rate, and to a lesser extent by how it handles errors such as aliasing

Introduction to ADC

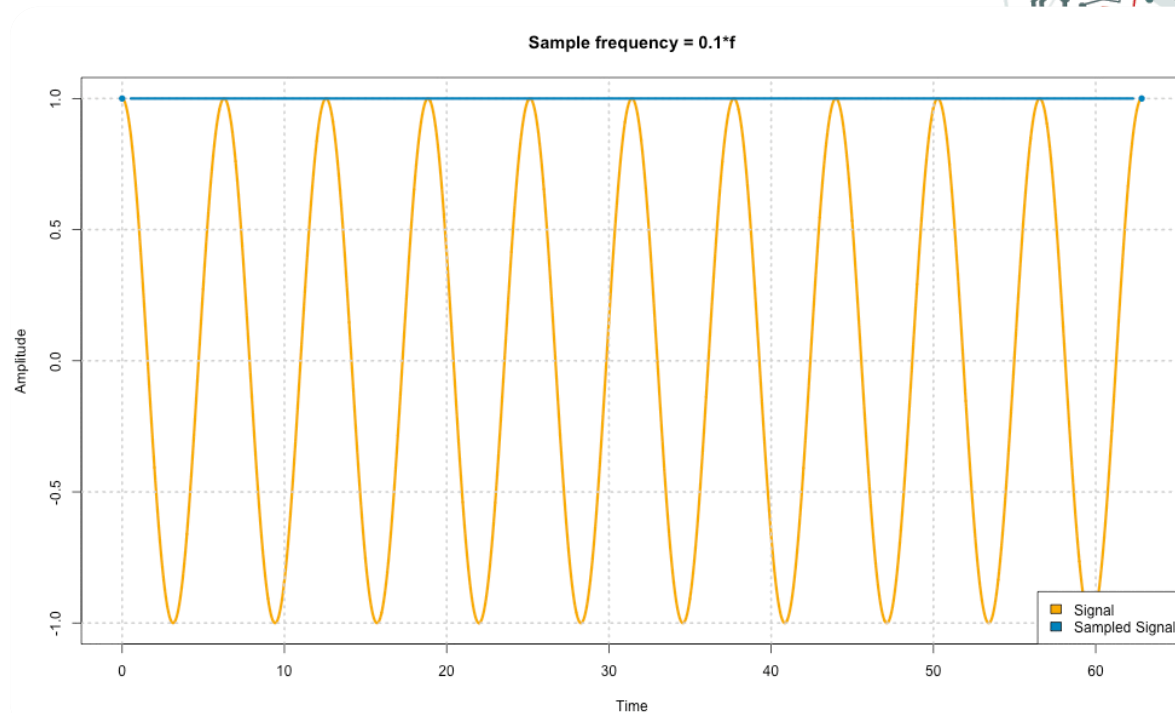
- More Common Types of ADC
 - Successive-approximation ADC (SAR)
 - Sigma-delta ADC
 - Pipeline



Introduction to ADC

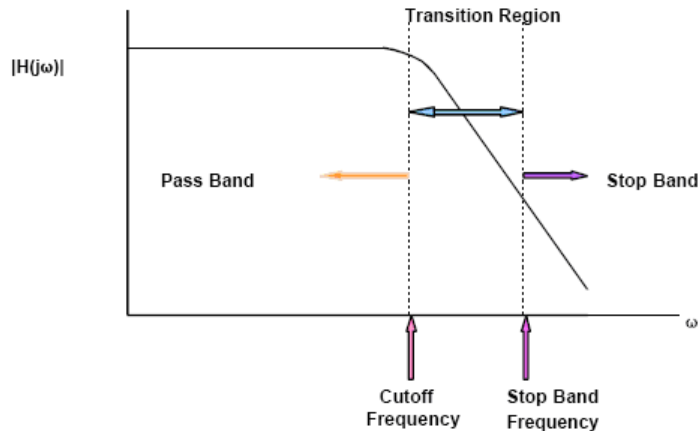
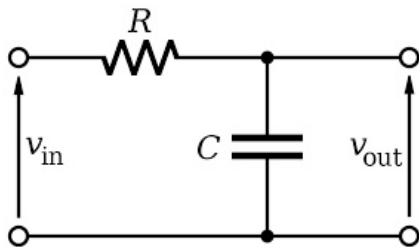
- Aliasing: Signal Distortion due to wrong sampling

Nyquist Rule: $f_s > 2f$



Introduction to ADC

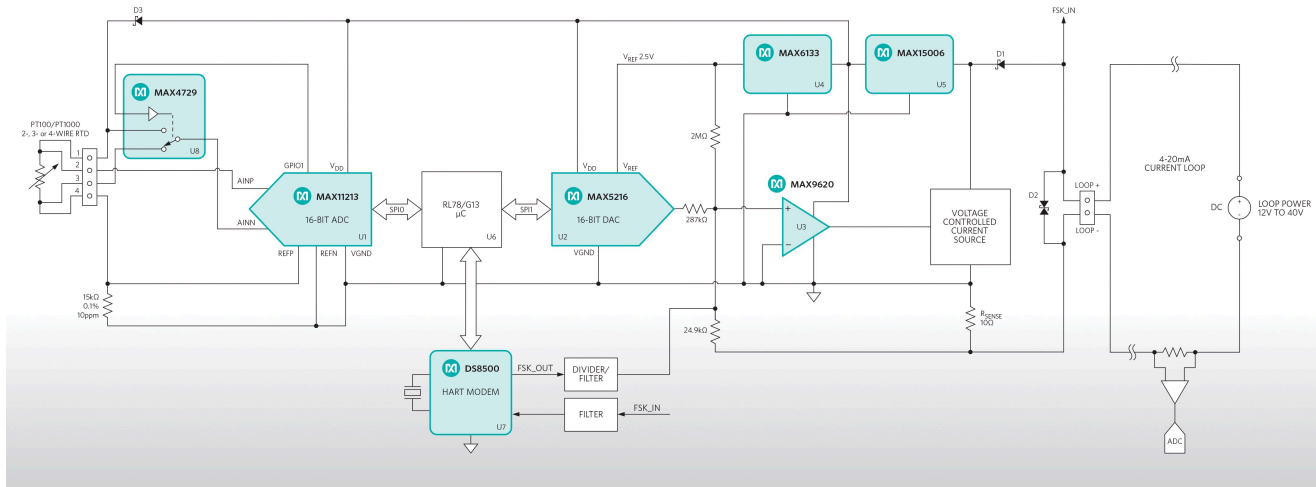
- Antia-Aliasing Filters
 - Cut the signal so it fits into the Nyquist Rule
 - Characterized by cut-off and stop-band frequencies



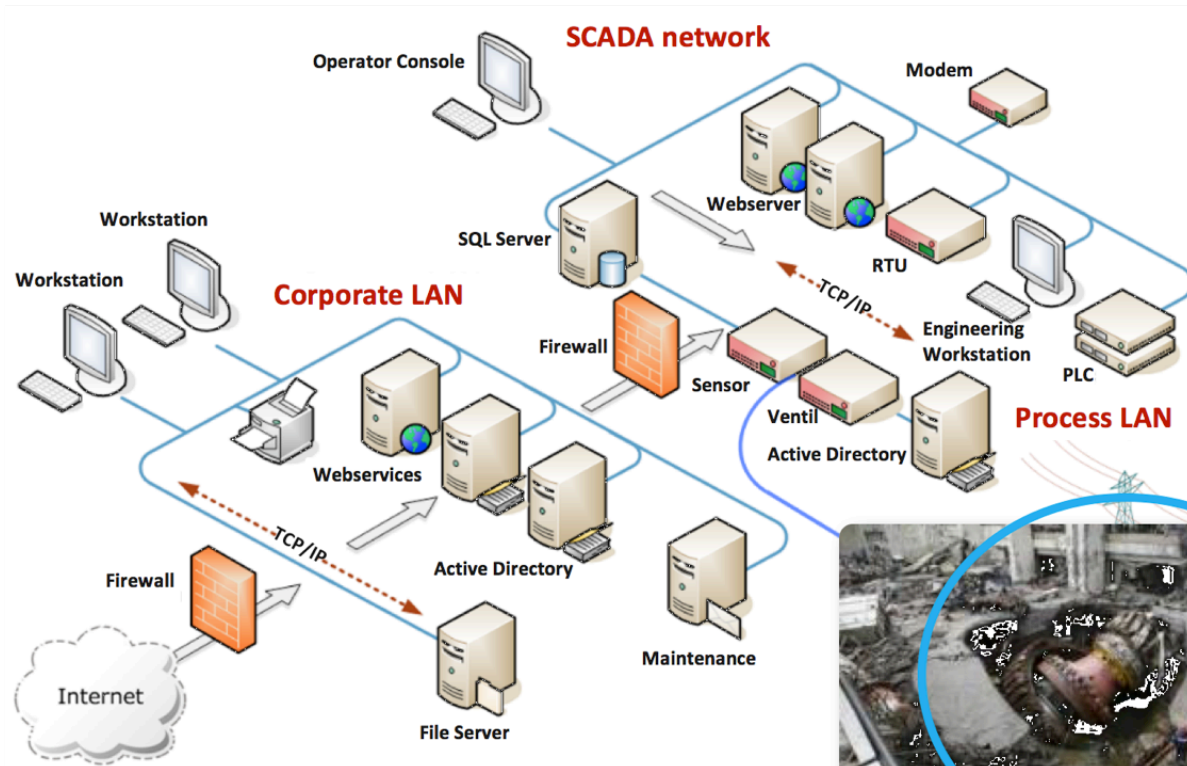
Introduction to ADC

- What is a DAC?

Novato Reference Design
MAXREFDES16#



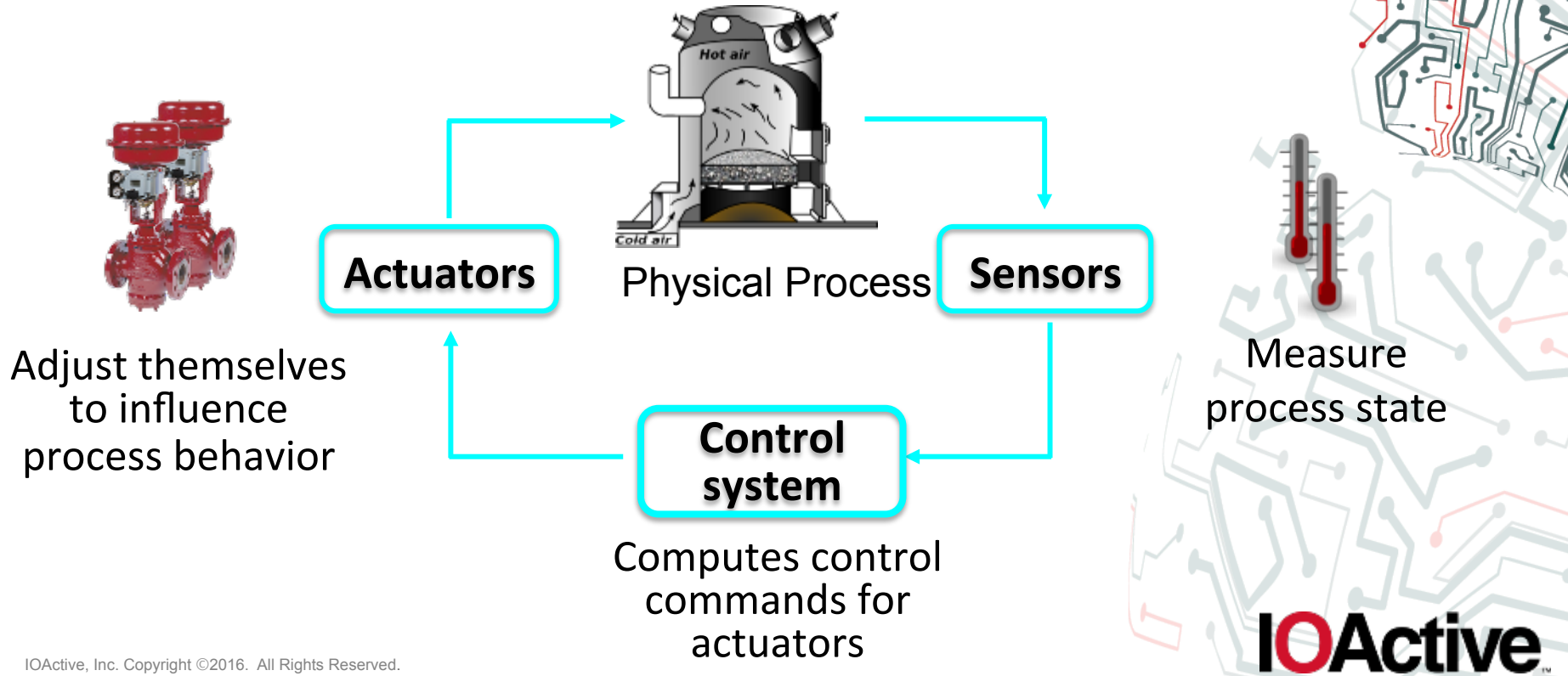
ADCs and ICSs



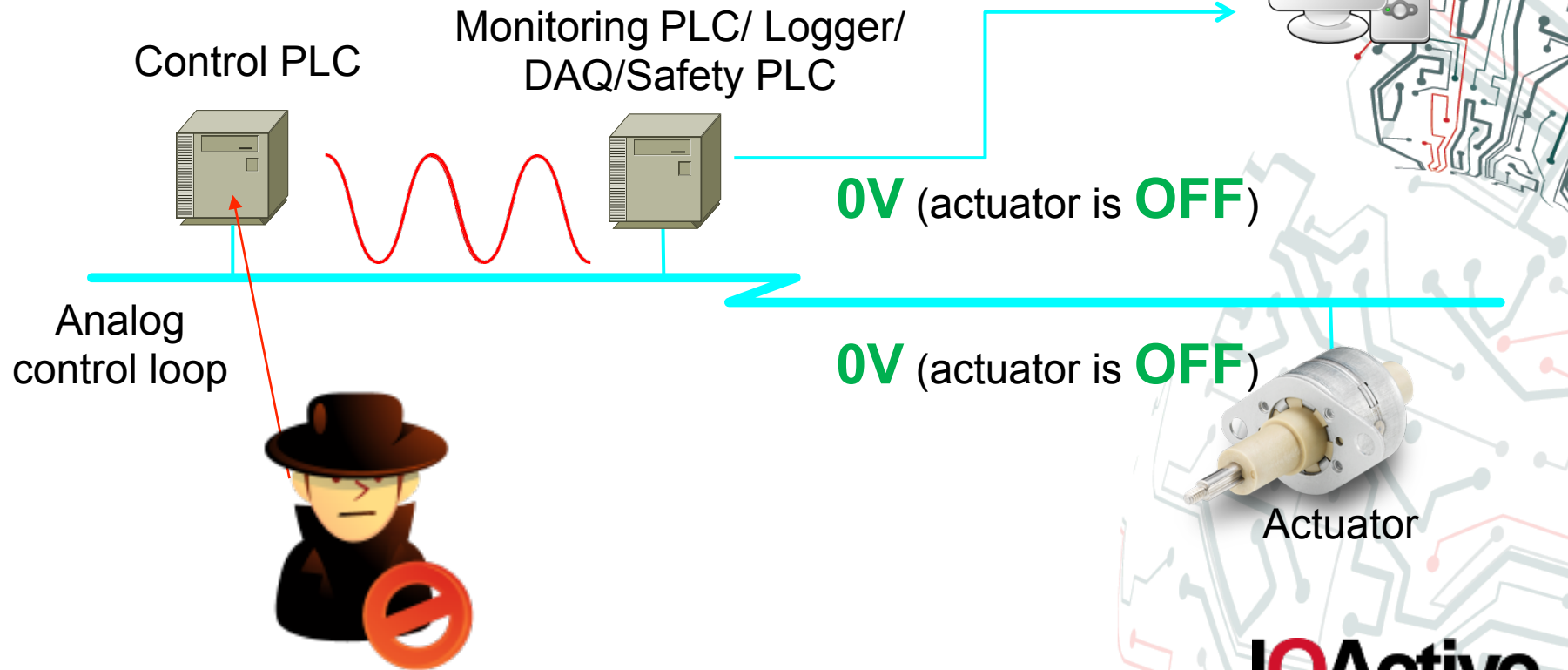
Catastrophic consequences



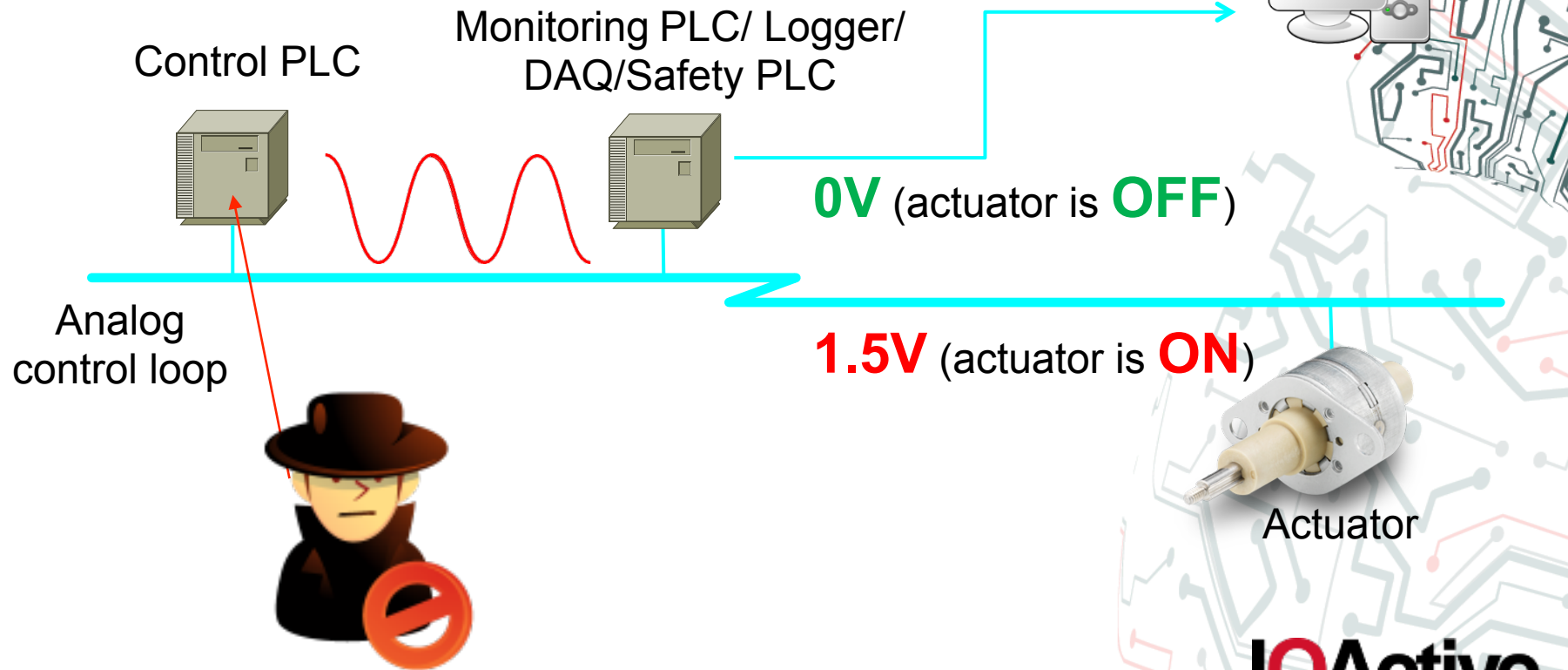
ADCs and ICSs



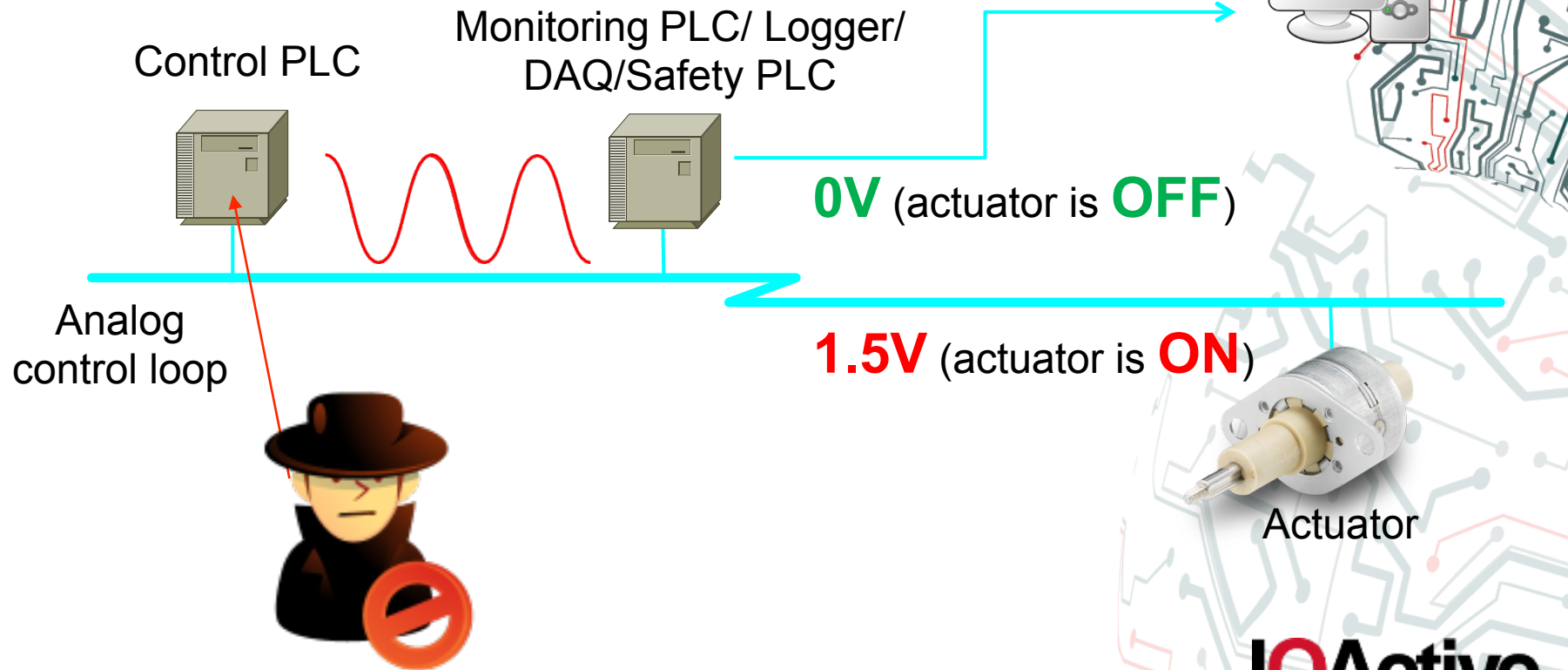
ADCs and ICSs



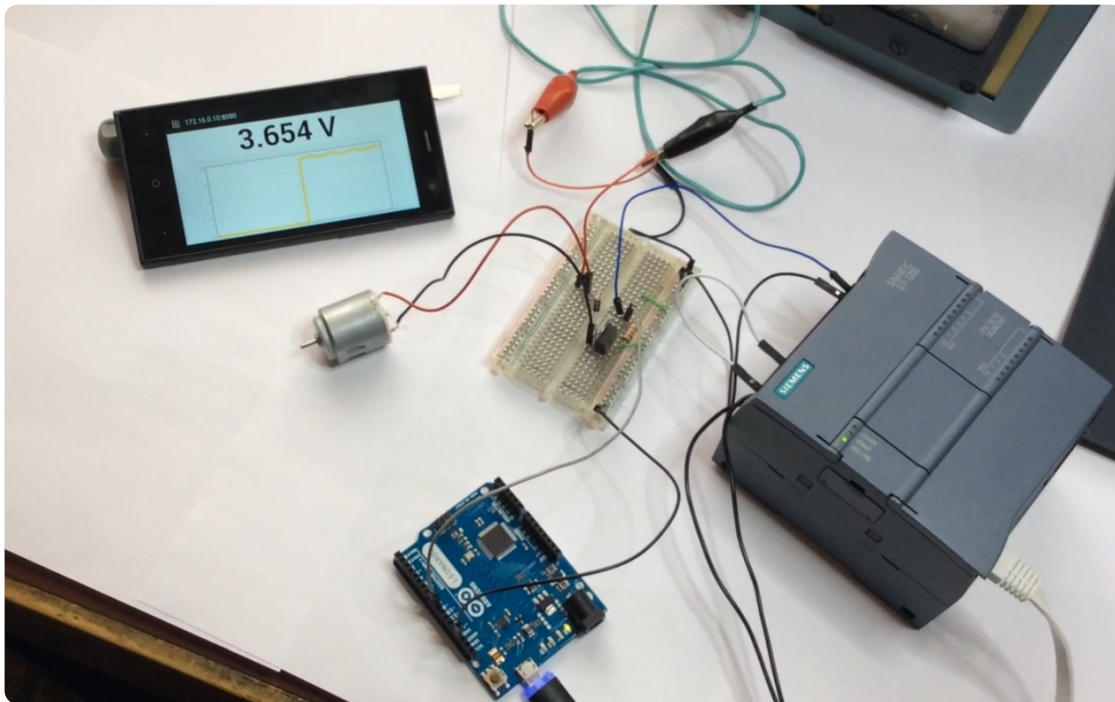
ADCs and ICSs



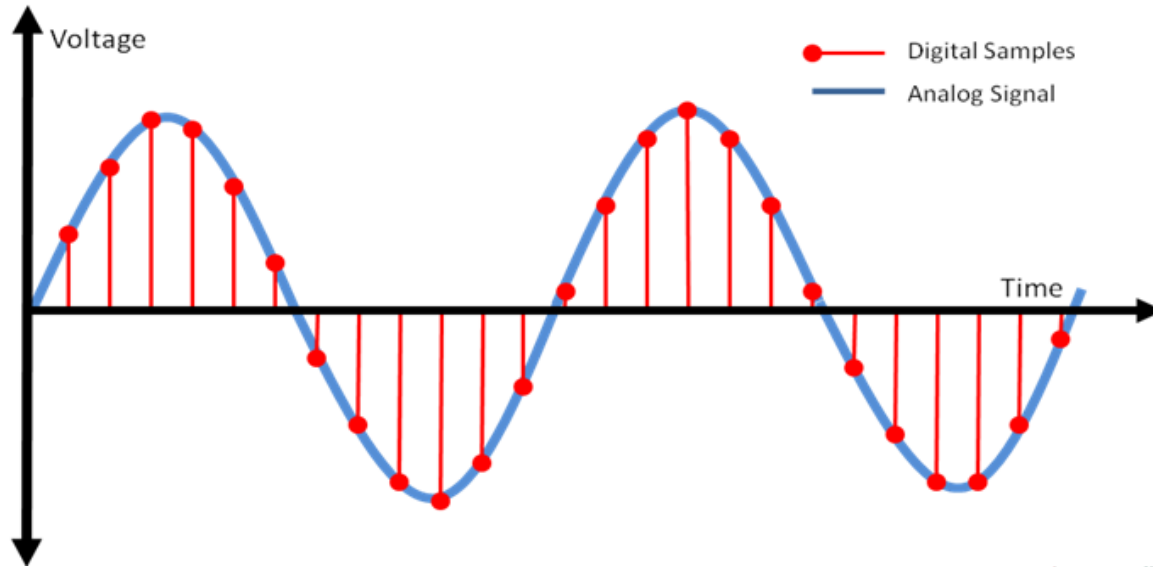
ADCs and ICSs



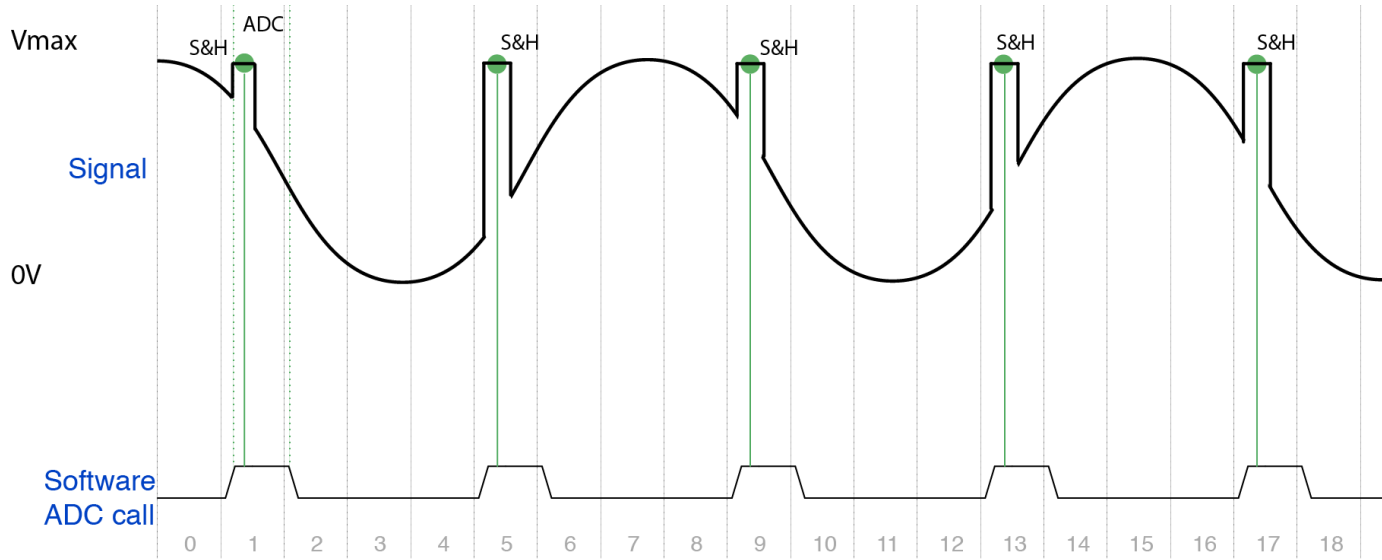
DEMO



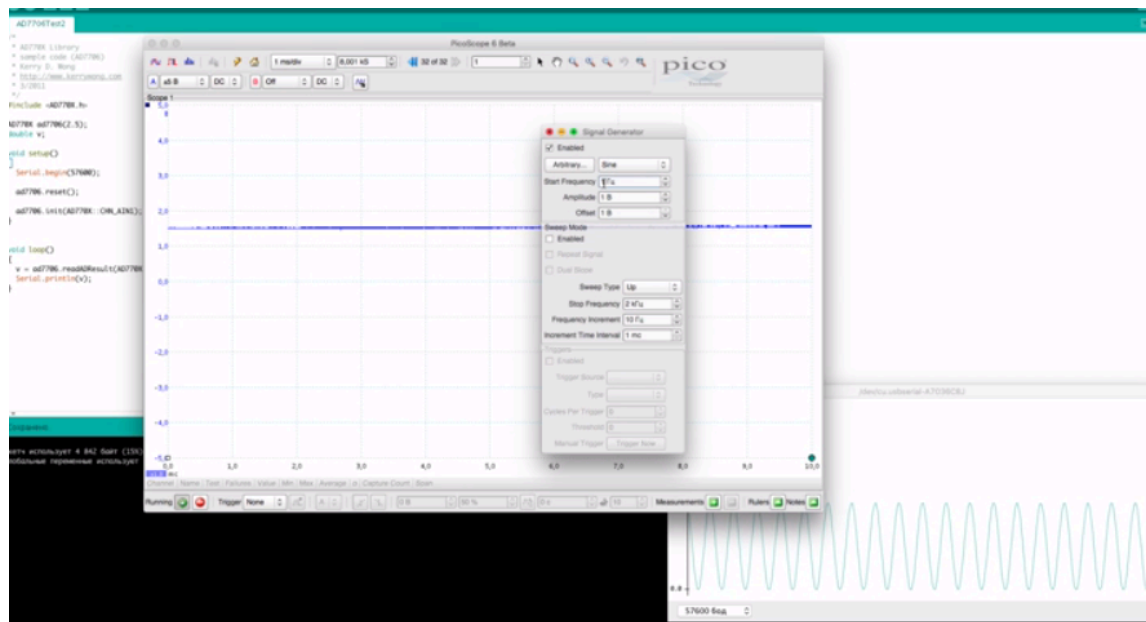
Attacking ADCs: Frequency



Attacking ADCs: Frequency

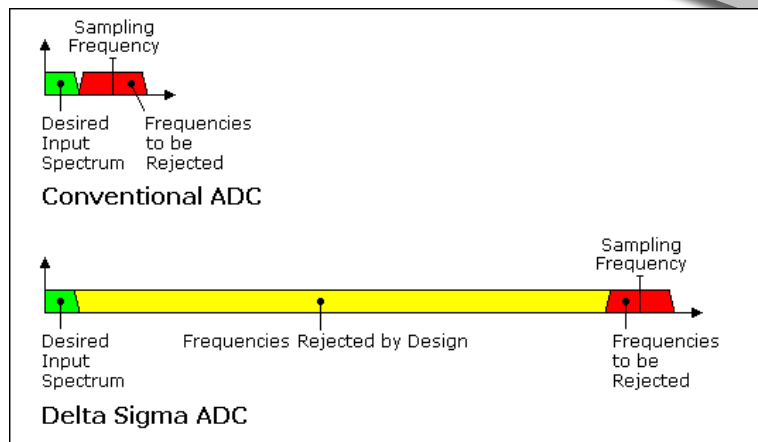


DEMO



Attacking ADCs: Frequency

- Delta – Sigma ADCs

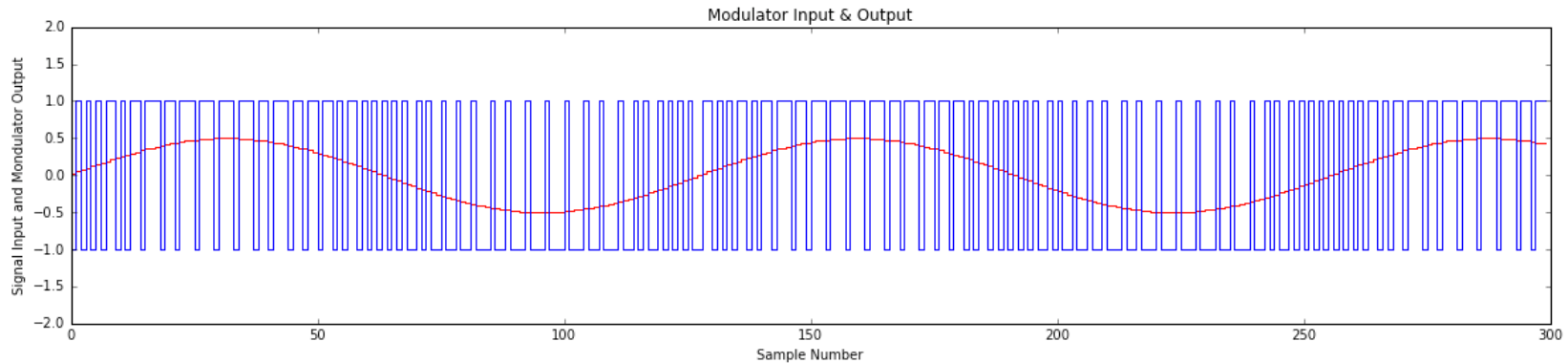


Thus it becomes obvious that conventional ADCs need expensive low pass filters in order to obtain a bandwidth close to the theoretical Nyquist limit. Delta sigma converters require simple RC low pass filters only and with a little more expense for a 2nd order filter one will get a virtually ideal behaviour. On the other hand an output low pass filter preceding the decimator is required, which again can be realized more precisely, easily and cheap in digital techniques. Note that in practice the proportions are much more extreme than in the graphic above as due to limited space an oversampling rate of approx. 16 only is shown there.

Oversampling
Delta-sigma ADCs use sample rates that are a large multiple, for instance, 128 times the sample rate sufficient for a given signal. For example, to sample a 25 kHz signal, a sample rate greater than the Nyquist rate (that is, > 50 KHz) would be sufficient. However, a delta-sigma ADC using an oversample factor of 128 samples the signal at 6 MHz. This approach has several benefits, such as better antialiasing and higher resolution.

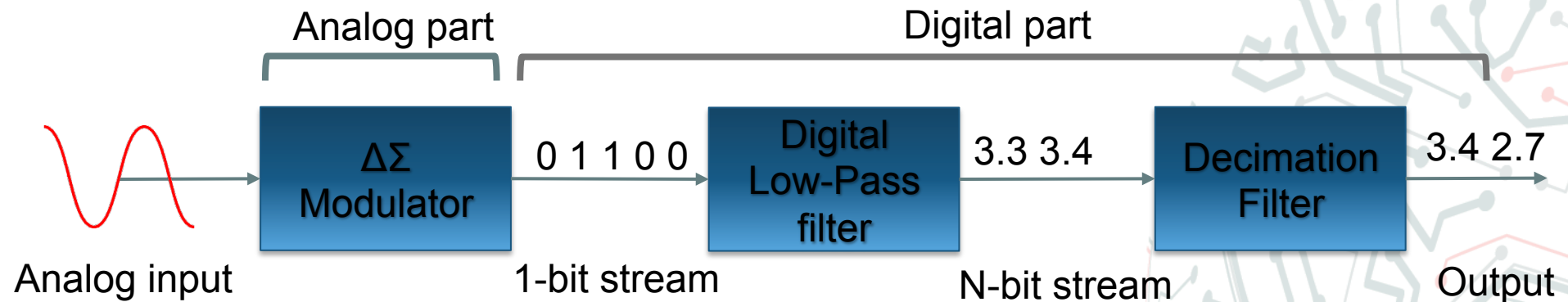
Attacking ADCs: Frequency

- Delta – Sigma ADCs
 - Δ -modulation the change in the signal is encoded.
 - The result is a stream of pulses
 - Accuracy improved using 1-bit DAC
 - Adding (Σ) the resulting analog signal to the input signal to reduce error



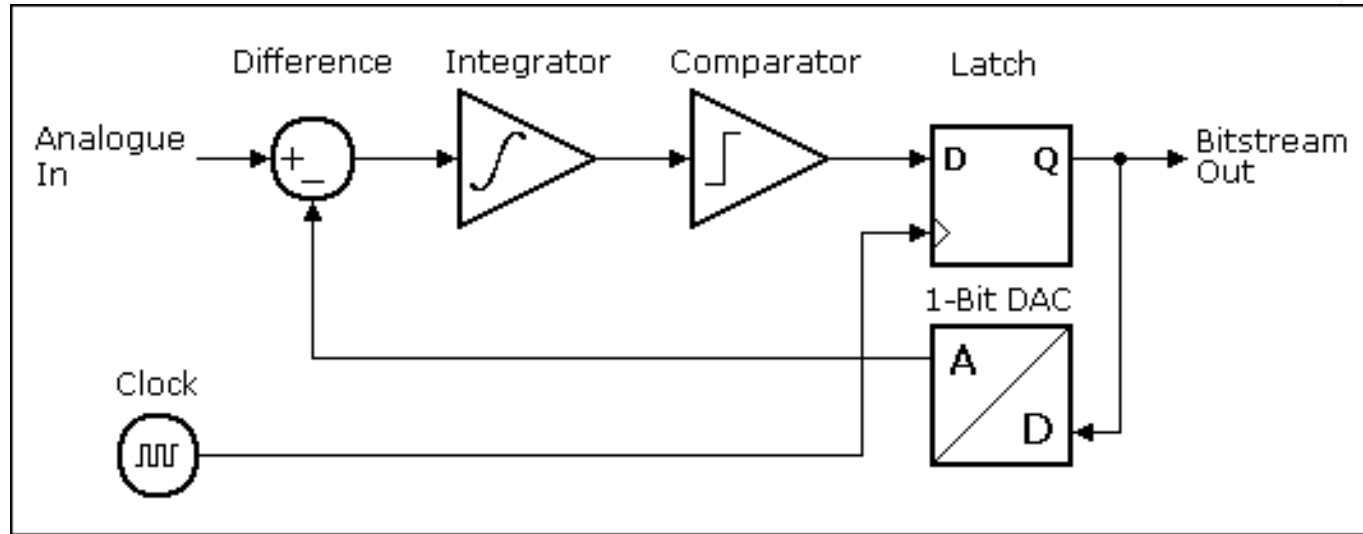
Attacking ADCs: Frequency

- Delta – Sigma ADCs
 - Analog part: High Frequency bitstream
 - Digital part: digital filter and decimation



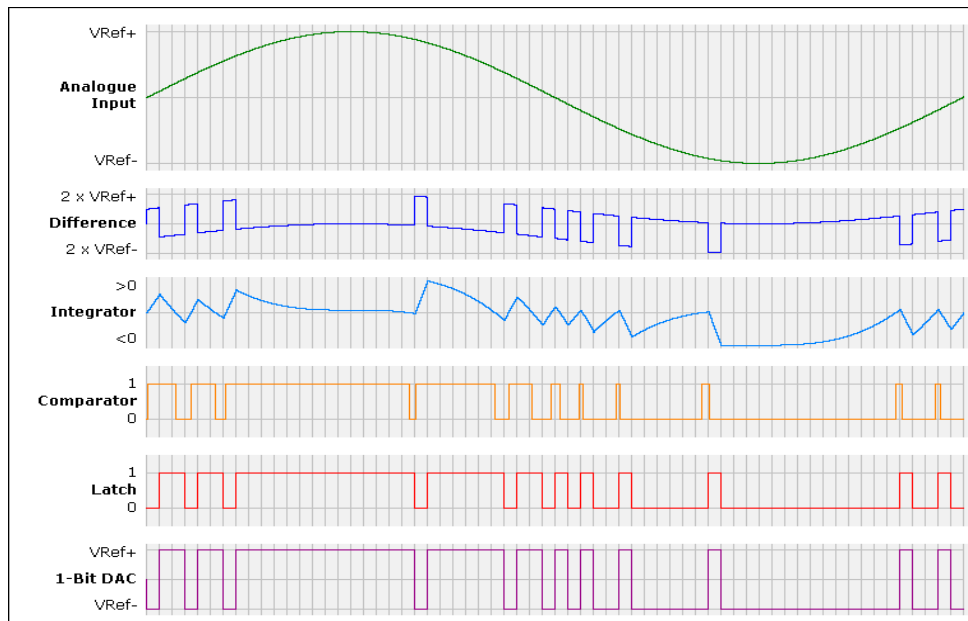
Attacking ADCs: Frequency

- Delta – Sigma ADCs
 - Analog part: High Frequency bitstream



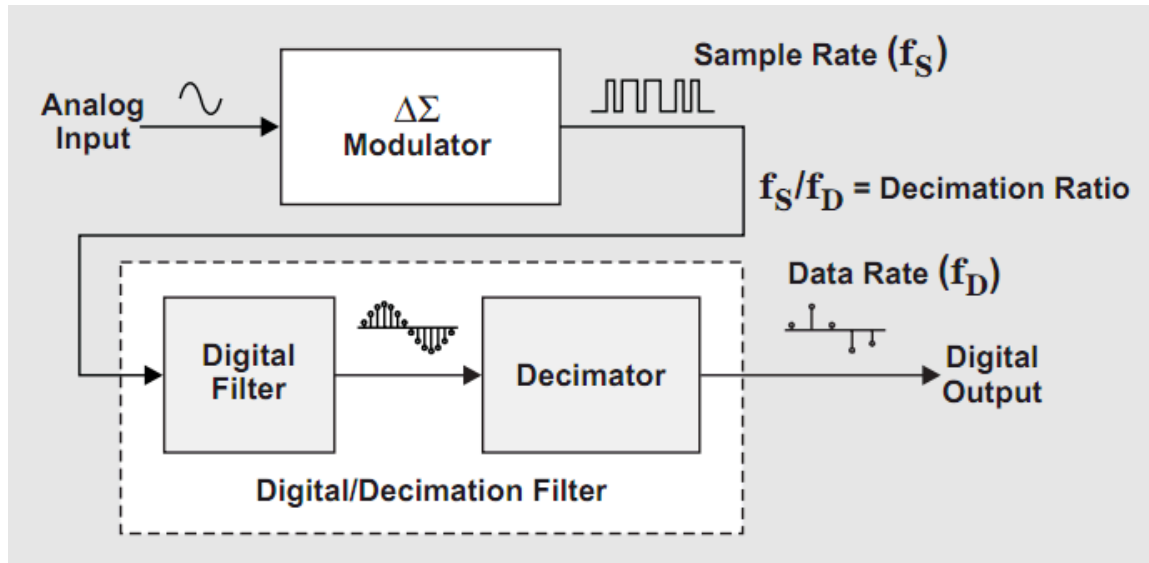
Attacking ADCs: Frequency

- Delta – Sigma ADCs
 - Modulation



Attacking ADCs: Frequency

- Delta – Sigma ADCs
 - Digital Part



Attacking ADCs: Frequency

- Delta – Sigma ADCs: Possible Problems
 - Filter Design / Implementation
 - Decimator Implementation

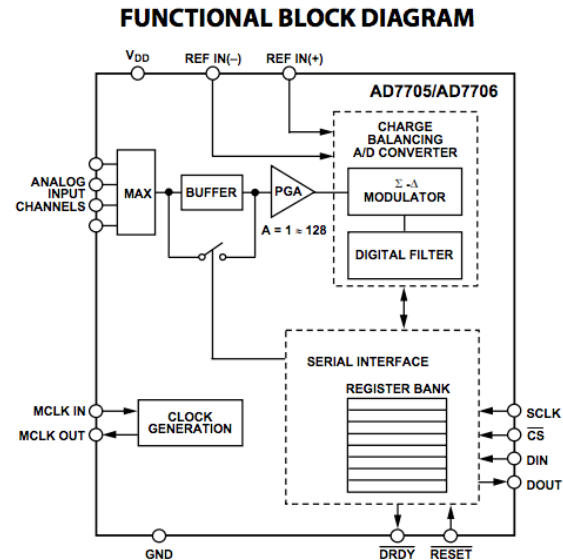
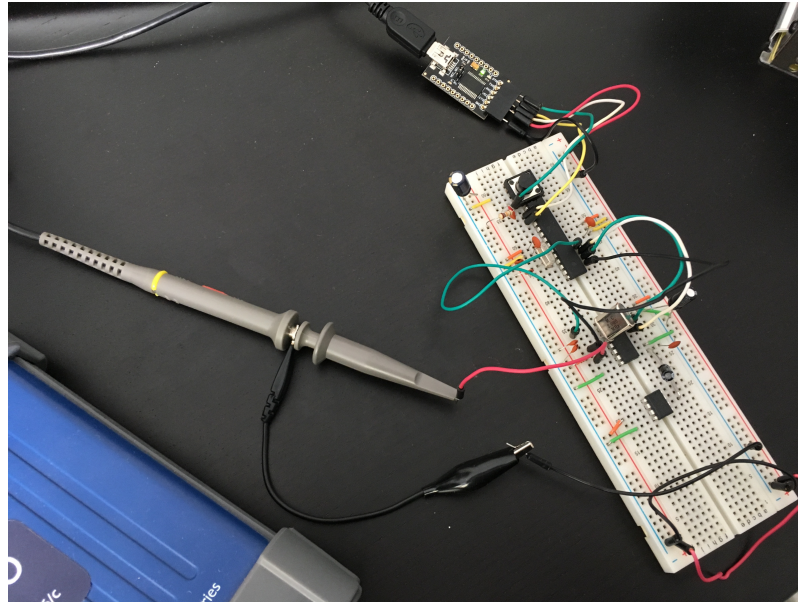


Figure 1.

Attacking ADCs: Frequency

- Delta – Sigma ADCs: Demo Setup

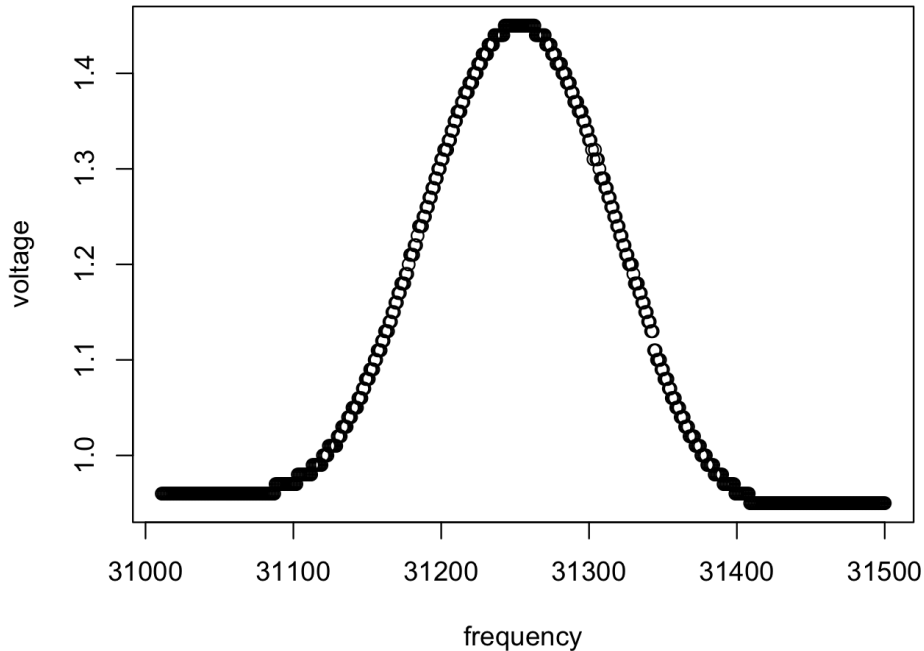


Attacking ADCs: Frequency

- Delta – Sigma ADCs: Demo Steps
 - Supply sinewave with frequency f to the ADC inputs
 - For 1 second acquire f_d voltage samples (equal to ADC samples per second rate) in digital form
 - Calculate and record $V_{\max}$, $V_{\min}$, V_{avg} and $\text{stddev}(V)$ for acquired samples
 - Increase frequency by 1 and goto (1)

Attacking ADCs: Frequency

- Delta – Sigma ADCs: Pardon me?



Attacking ADCs: Frequency

- Delta – Sigma ADCs: Results

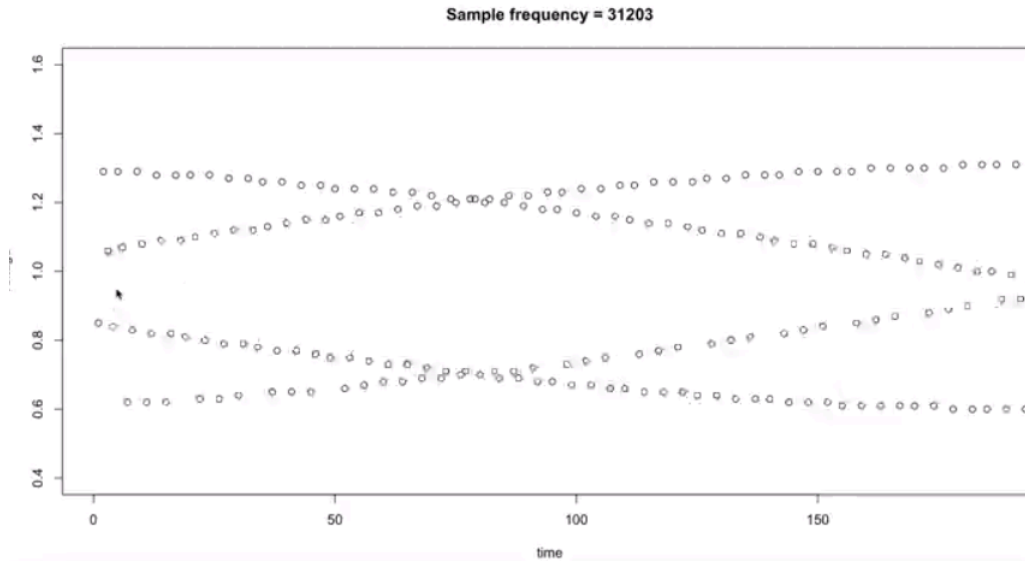
Table 23. Input Sampling Frequency vs. Gain

Gain	Input Sampling Frequency (f_s)
1	$f_{CLKIN}/64$ (38.4 kHz @ $f_{CLKIN} = 2.4576$ MHz)
2	$2 \times f_{CLKIN}/64$ (76.8 kHz @ $f_{CLKIN} = 2.4576$ MHz)
4	$4 \times f_{CLKIN}/64$ (153.6 kHz @ $f_{CLKIN} = 2.4576$ MHz)
8 to 128	$8 \times f_{CLKIN}/64$ (307.2 kHz @ $f_{CLKIN} = 2.4576$ MHz)

In addition, the digital filter does not provide any rejection at integer multiples of the digital filter's sample frequency. However, the input sampling on the part provides attenuation at multiples of the digital filter's sampling frequency so that the unattenuated bands occur around multiples of the sampling frequency, f_s , as defined in Table 23. Thus, the unattenuated bands occur at $n \times f_s$ (where $n = 1, 2, 3 \dots$). At these frequencies, there are frequency bands $\pm f_{3\text{ dB}}$ wide ($f_{3\text{ dB}}$ is the cutoff frequency of the digital filter) at either side where noise passes unattenuated to the output.

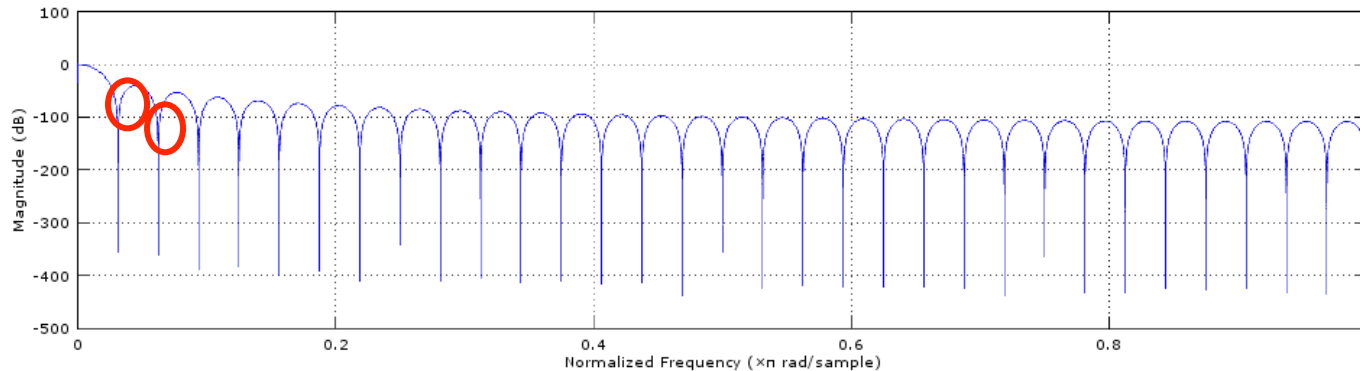
Attacking ADCs: Frequency

- Delta – Sigma ADCs: DEMO



Attacking ADCs: Frequency

- Delta – Sigma ADCs: Possible Explanation
 - This areas could case some noise transition, but not of that size

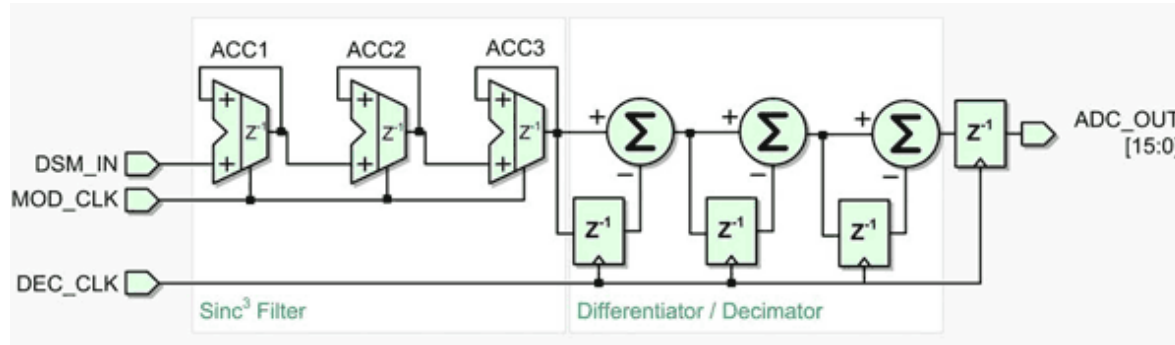


Attacking ADCs: Frequency

- Delta – Sigma ADCs: Possible Explanation
 - Could it be incorrect implementation of digital LPF? But incorrect in what?
 - If (IF!) our filter is implemented badly, we could be in a situation with simple integer overflow.
 - Probability of last statement arises if we have "MCU" inside ADC

Attacking ADCs: Frequency

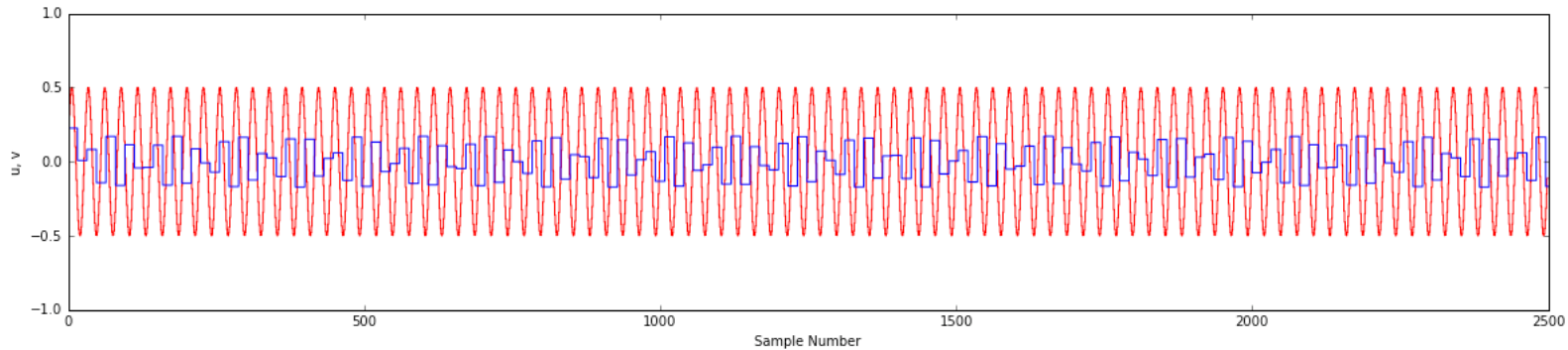
- Delta – Sigma ADCs: Possible Explanation



```
// apply the filter to each input sample
for ( n = 0; n < length; n++ ) {
    // calculate output n
    coeffp = coeffs;
    inputp = &insamp[filterLength - 1 + n];
    acc = 0;
    for ( k = 0; k < filterLength; k++ ) {
        acc += (*coeffp++) * (*inputp--);
    }
    output[n] = acc;
}
```

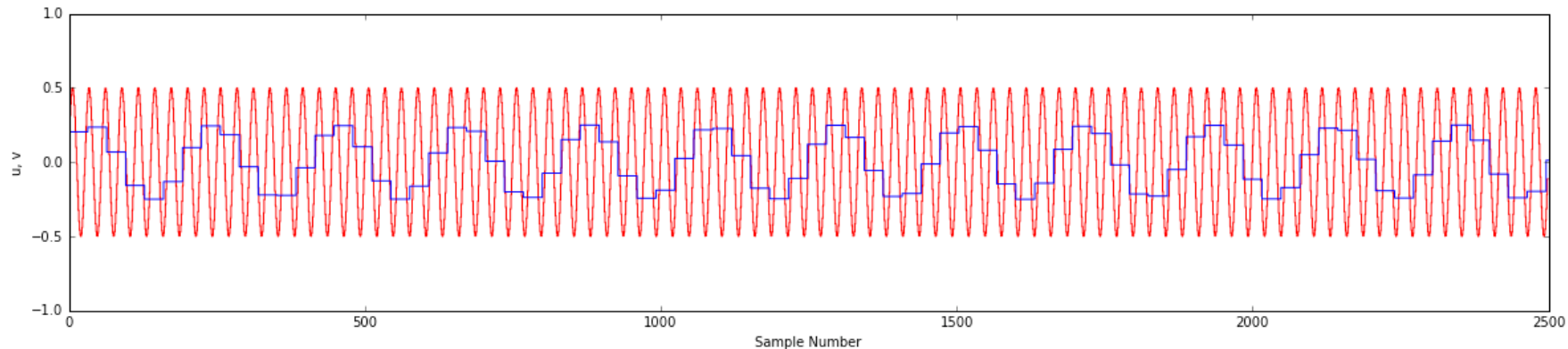
Attacking ADCs: Frequency

- Delta – Sigma ADCs: Possible Explanation



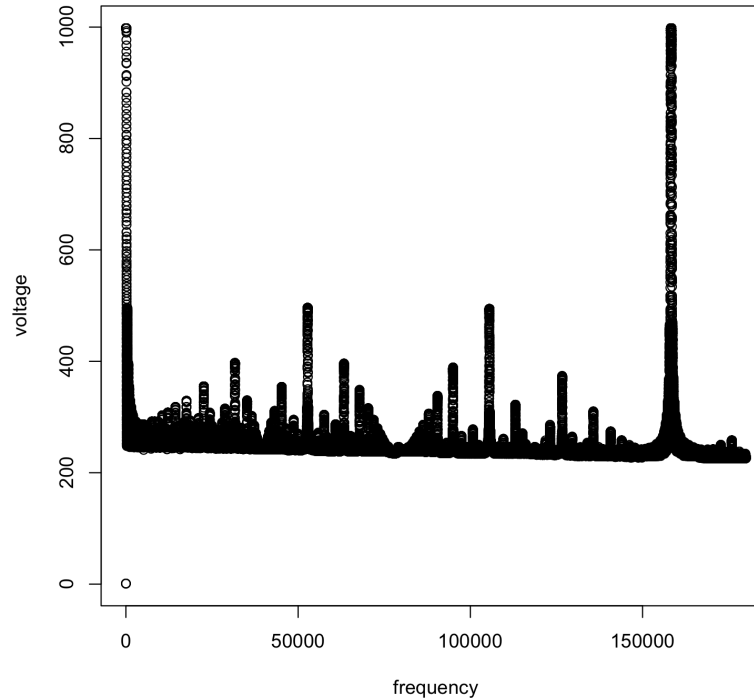
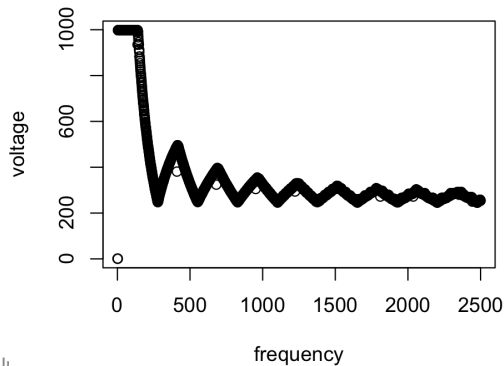
Attacking ADCs: Frequency

- Delta – Sigma ADCs: Possible Explanation



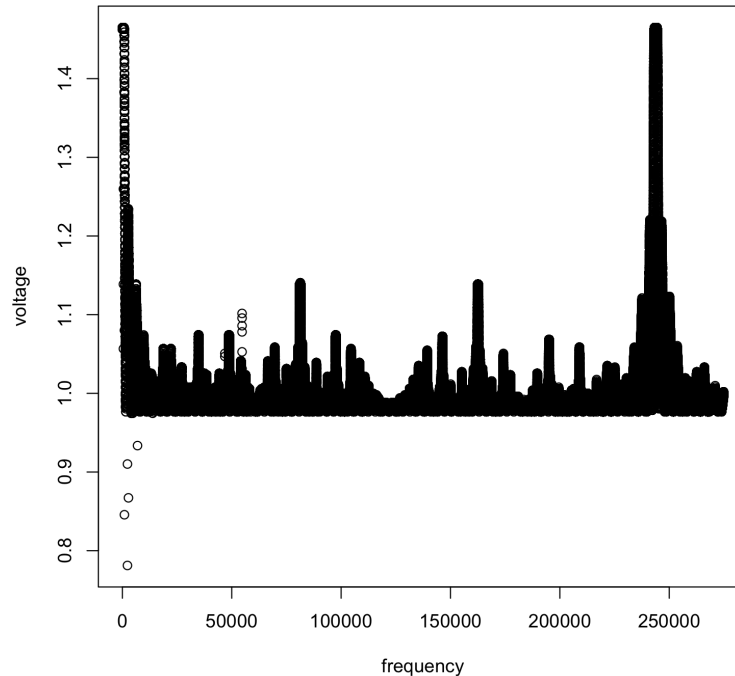
Attacking ADCs: Frequency

- Delta – Sigma ADCs: MCP3425



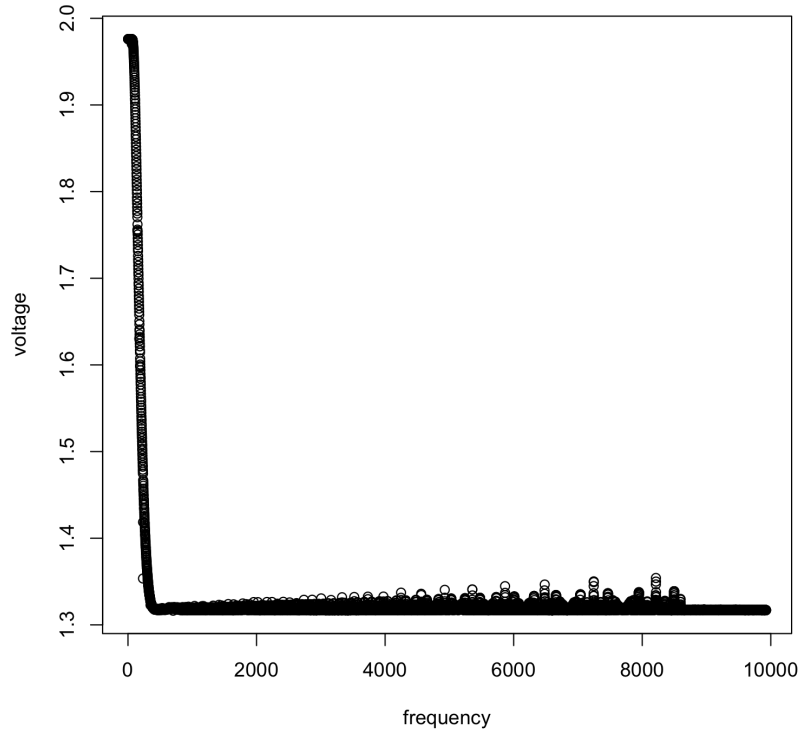
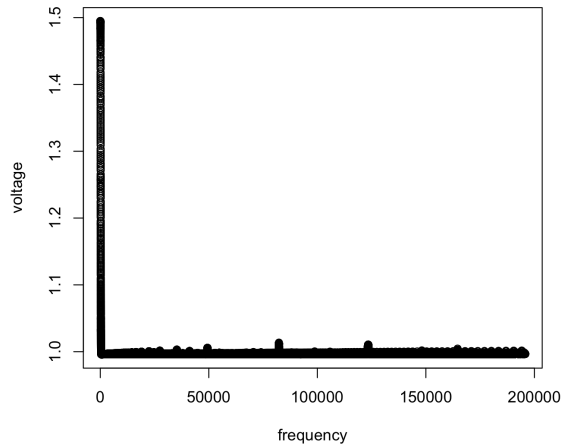
Attacking ADCs: Frequency

- Delta – Sigma ADCs: ADS1015



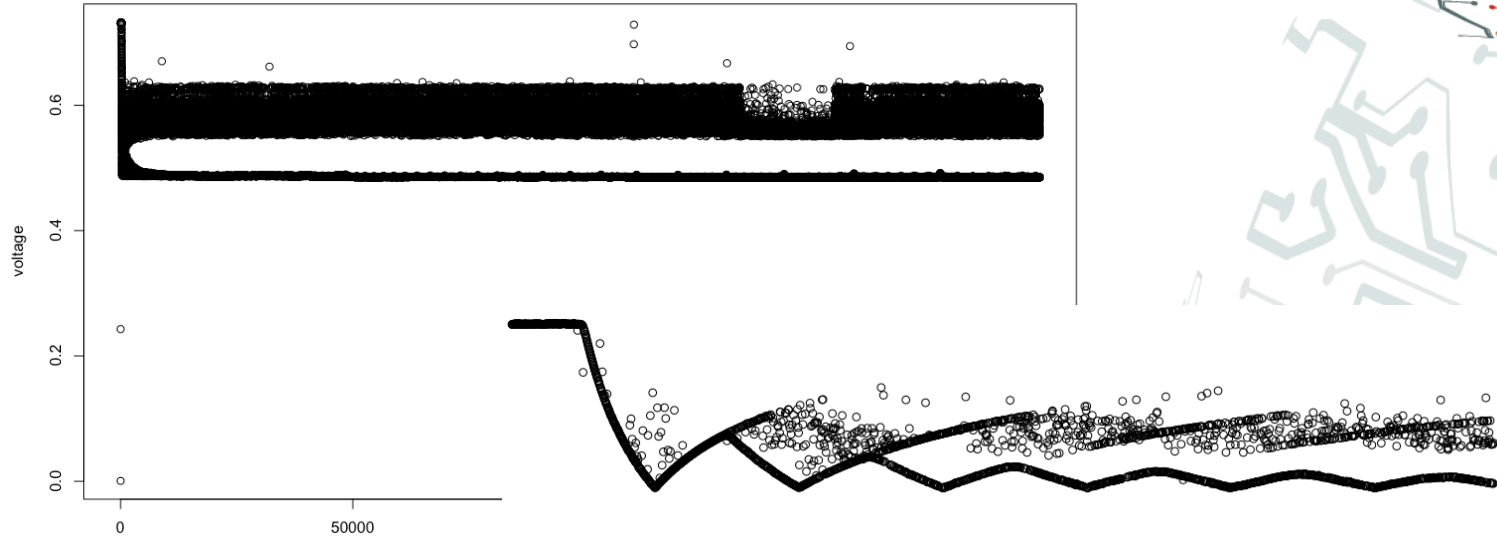
Attacking ADCs: Frequency

- Delta – Sigma ADCs: MAX11205



Attacking ADCs: Frequency

- Delta – Sigma ADCs: Cypress PSoC5

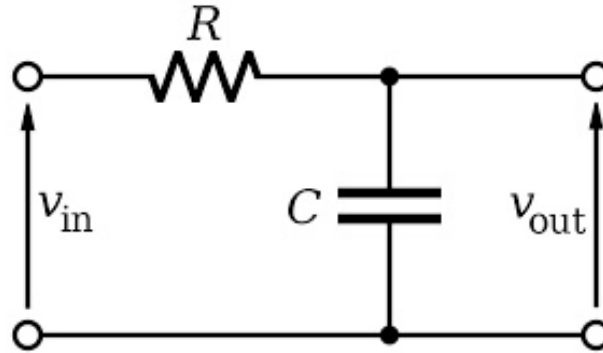


Attacking ADCs: Frequency

- Delta – Sigma ADCs: Summary

ADC	First "attackable" f	Required AAF f_c	Required AAF f_{sb}	Attack complexity
AD7705/AD7706	31250/38400 Hz	-	30kHz	easy
MCP3425	~51kHz	10-20kHz	30kHz	easy/medium
ADS1015	~86kHz	10-20kHz	50kHz	medium/hard
MAX11205	n/a	any reasonable	any reasonable	~impossible
PSoC5 LP*	~1kHz	1kHz	2kHz	n/a (medium)

Conclusion



- Not all $\Delta\Sigma$ are the same
- Use Anti Aliasing Filter before the Input

Q & A